

High performance AlGaN/GaN HEMTs with 2.4 μm source–drain spacing*

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Abstract: This paper describes the performance of AlGaN/GaN HEMTs with 2.4 μm source–drain spacing. So far these are the smallest source–drain spacing AlGaN/GaN HEMTs which have been implemented with a domestic wafer and domestic process. This paper also compares their performance with that of 4 μm source–drain spacing devices. The former exhibit higher drain current, higher gain, and higher efficiency. It is especially significant that the maximum frequency of oscillation noticeably increased.

Key words: AlGaN/GaN HEMT; high frequency; source–drain spacing

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1. Introduction

Wide-bandgap AlGaN/GaN high-electron-mobility transistors (HEMTs) are being developed for high power and high frequency applications. This is because GaN-based materials exhibit high current density, high saturation velocity and high breakdown field. 40 W/mm power operation has been demonstrated at 4 GHz^[1]. There is also an increasing interest in applications at higher frequencies. Ka-band and even W-band operations have been reported abroad^[2–4].

An important feature of most millimeter-wave devices is that they have small source–drain spacings (L_{sd}). Some studies have indicated that the gate–drain spacing (L_{gd}) can influence the maximum frequency of oscillation (f_{max})^[3] and that reducing the gate–source spacing (L_{gs}) can increase drain current and transconductance^[5]. Yet there have been no general comparisons between different source–drain spacings. On the other hand, reducing L_{sd} is believed to be difficult. The most important reason is that the conventional Ti/Al/Ti/Au ohmic contact has rough morphology and edges after annealing, which limits the reduction of the spacing. At present, most domestic AlGaN/GaN HEMTs have L_{sd} of over 4 μm .

According to our study, the ohmic contact edges can be improved by using Ti/Al/Ni/Au. Being a more effective diffusion barrier, nickel (Ni) can prevent the formation of an Al–Au alloy, which can scatter along the wafer surface and causes rough morphology and edges^[6]. Moreover, the ohmic contact resistivity of Ti/Al/Ni/Au can reach $10^{-6} \Omega\cdot\text{cm}^2$ which is equivalent to that of Ti/Al/Ti/Au.

Based on this study, we fabricated AlGaN/GaN HEMTs with L_{sd} of 2.4 μm . To the best of our knowledge, so far these are the smallest source–drain spacing devices implemented with a domestic wafer and process. We also compare their performance with that of 4 μm source–drain spacing devices on the same wafer. The 2.4 μm source–drain spacing devices exhibit better performance.

2. Device structure and fabrication

The two-inch epitaxial wafer was provided by the Institute of Semiconductors of the Chinese Academy of Sciences. Al_{0.25}Ga_{0.75}N/AlN/GaN multilayers were grown on a sapphire substrate by MOCVD and the sheet resistance is 354 Ω/\square . Ohmic metal consisting of Ti/Al/Ni/Au was deposited by evaporation, followed by rapid thermal annealing at 870 °C for 50 s. The ohmic contact resistivity was $8.7 \times 10^{-6} \Omega\cdot\text{cm}^2$ measured by the 4-probe method. Passivation was done by using SiN film (120 nm) deposited by PECVD. The T-shape Ni/Au gate was formed by the combined processes of electron-beam lithography, dry-etch and evaporation. The completed gate length is 0.35 μm and the length of the gate head is 0.7 μm . The unit gate width is 75 μm . The gate–source spacing is 0.7 μm for the SD 2.4 μm devices, and is 1.5 μm for the SD 4 μm devices. Figure 1 shows a picture of an SD 2.4 μm device.

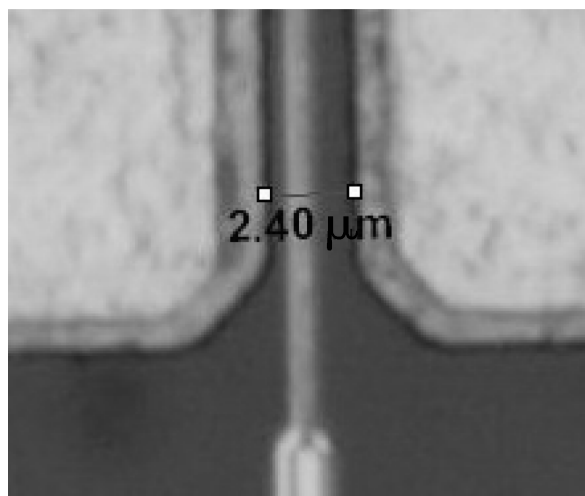


Fig. 1. An AlGaN/GaN HEMT with 2.4 μm source–drain spacing.

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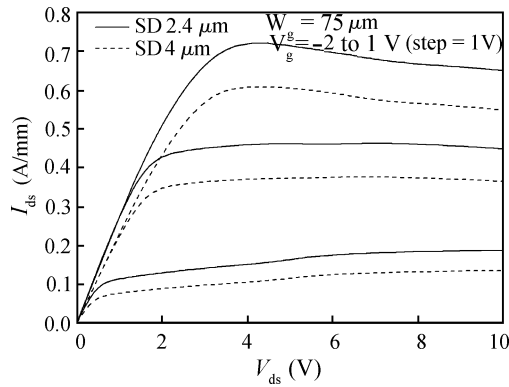


Fig. 2. Drain I - V characteristics of devices with SD $2.4\ \mu\text{m}$ and SD $4\ \mu\text{m}$ respectively.

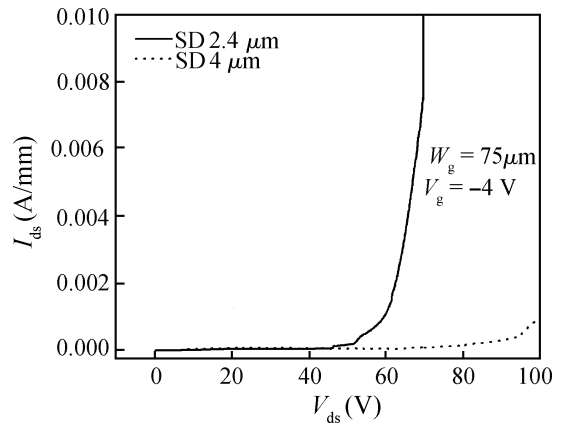


Fig. 4. Breakdown characteristics of devices with SD $2.4\ \mu\text{m}$ and SD $4\ \mu\text{m}$ respectively.

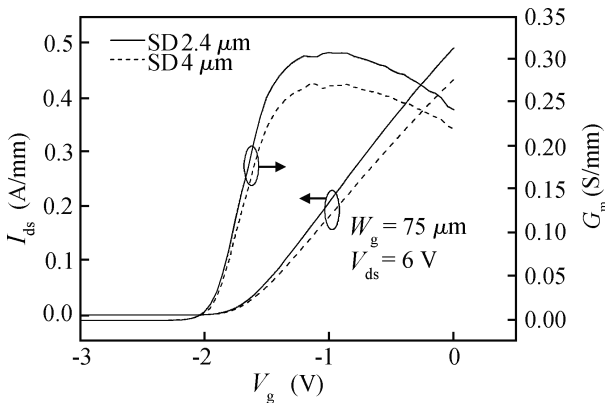


Fig. 3. Transconductance (g_m) and I_{ds} - V_g characteristics of devices with SD $2.4\ \mu\text{m}$ and SD $4\ \mu\text{m}$ respectively.

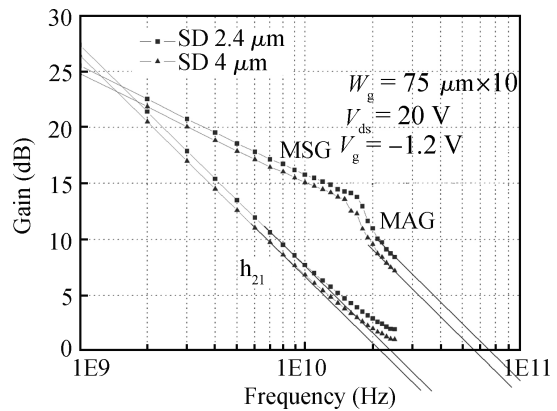


Fig. 5. Small-signal RF gains of devices with SD $2.4\ \mu\text{m}$ and SD $4\ \mu\text{m}$ respectively.

3. DC and small-signal characteristics

Figure 2 shows the drain I - V characteristics for devices with SD $2.4\ \mu\text{m}$ and SD $4\ \mu\text{m}$ respectively. Typically, the SD $4\ \mu\text{m}$ devices exhibit a maximum drain current (I_{DSS}) of 608 mA/mm, and the knee voltage (V_k) is 2.80 V at $V_g = 1\ \text{V}$. For the SD $2.4\ \mu\text{m}$ devices, the drain current is 721 mA/mm and the V_k is 2.86 V. The drain current of the latter is 18% higher, yet the knee voltage is similar.

Figure 3 shows the transconductance (g_m) and I_{ds} - V_g characteristics of the two kinds of devices at $V_{ds} = 6\ \text{V}$. The maximum g_m of the SD $2.4\ \mu\text{m}$ device is 308 mS/mm, and the SD $4\ \mu\text{m}$ one's is 272 mS/mm. The increased g_m of the former is due to the reduction of the gate-source spacing^[3]. Figure 4 shows the breakdown characteristics of the two kinds of devices. If taking the breakdown condition as the drain current reaching 1 mA/mm, the breakdown voltages of the SD $2.4\ \mu\text{m}$ and the SD $4\ \mu\text{m}$ devices are 62 V and 100 V respectively. Considering that an operating voltage of 28 V is widely used in microwave systems, 62 V is enough for such applications.

Figure 5 shows the current gain (h_{21}) and MSG/MAG (maximum stable gain/maximum available gain) of the two kinds of devices. Both the devices have a gate width of $75\ \mu\text{m} \times 10$. Table 1 shows a comparison of the small signal RF gains of the two devices.

It can be seen that the SD $2.4\ \mu\text{m}$ device exhibits higher frequency and higher gain characteristics. Also, the SD $2.4\ \mu\text{m}$

Table 1. Comparison of RF gains for devices with SD $2.4\ \mu\text{m}$ and SD $4\ \mu\text{m}$ respectively.

Parameter	SD $4\ \mu\text{m}$	SD $2.4\ \mu\text{m}$
f_T (GHz)	22	24
f_{max} (GHz)	57	67
MSG at 8 GHz (dB)	16.0	16.7
MAG at 30 GHz (dB)	5.7	7.0
Turning frequency from MSG to MAG (GHz)	15	17

device has a higher turning frequency from MSG to MAG, which is helpful to increase the gain at high frequency. At 30 GHz, the maximum gain of the SD $2.4\ \mu\text{m}$ device is 1.3 dB higher. This indicates that reducing the source-drain spacing is important for high frequency HEMTs.

By reducing the gate-source spacing and gate-drain spacing, the series resistances R_S and R_D are lowered. So the SD $2.4\ \mu\text{m}$ devices exhibit better DC and small-signal characteristics as above. Also, because of the reduction of the gate-drain spacing, the breakdown voltage is lowered.

4. Power operation

Figure 5 shows the large-signal RF performance of the two kinds of devices with a gate width of $75\ \mu\text{m} \times 10$. The devices

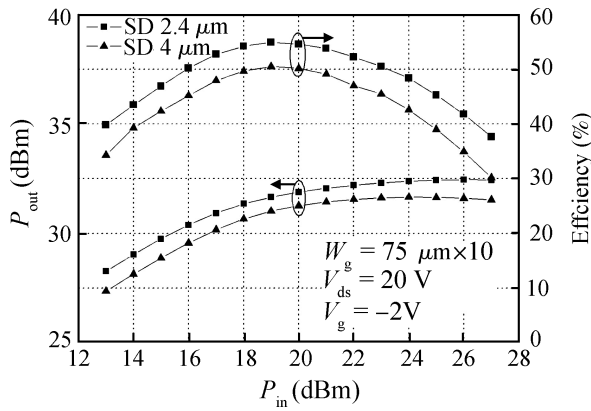


Fig. 6. Output power and PAE of devices with SD 2.4 μm and SD 4 μm respectively at 8 GHz.

operated in deep AB class at $V_{ds} = 20 \text{ V}$, and were measured on-wafer with a load-pull system at 8 GHz. The SD 2.4 μm device has a saturated output power of 32.4 dBm, which is 0.8 dB (20%) higher than that of the SD 4 μm device. Moreover, the gain and the PAE (power added efficiency) are also higher.

The higher output power of the SD 2.4 μm device is due to its higher I_{dss} and relatively low V_k . Because the gate-source spacing and gate-drain spacing are reduced, the power consumed at the resistances R_S and R_D decreases, and the PAE increases.

5. Conclusion

We have successfully fabricated AlGaIn/GaN HEMTs with 2.4 μm source-drain spacing. So far these are the smallest source-drain spacing devices which have been implemented with a domestic wafer and process. Compared with the 4 μm source-drain spacing devices, the former exhibit higher drain current, higher gain, and higher efficiency. In particular, the maximum frequency of oscillation is noticeably enhanced.

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