

# Alloy Temperature Dependence of Offset Voltage and Ohmic Contact Resistance in Thin Base InGaP/GaAs HBTs\*

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**Abstract:** The alloy temperature dependence of  $V_{\text{offset}}$  and  $R_{\text{contact}}$  is studied, and an optimal alloy temperature range for the best trade-off between  $V_{\text{offset}}$  and  $R_{\text{contact}}$  is given for thin base HBTs. In addition, the reason for the high  $V_{\text{offset}}$  at high alloy temperature is interpreted using Schottky clamped theory. The lower  $V_{\text{offset}}$  of our U-shaped emitter HBT than that of traditional strip emitter HBTs is explained.

**Key words:** heterojunction bipolar transistor; U-shaped emitter; alloy; offset voltage

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## 1 Introduction

InGaP/GaAs HBTs have inherent advantages, including low  $1/f$  noise, low manufacturing cost, and a reliable fabrication process, that make them attractive for use in MMIC circuits<sup>[1-5]</sup>. The cutoff frequency  $f_t$  of a device is a key parameter that directly confines circuit operation speed<sup>[6]</sup>. Decreasing the base thickness is an effective way to enhance  $f_t$ . In the fabrication of HBTs, a high-temperature alloy process is usually adopted to form an ohmic contact to the collector. It is necessary to consider the influence of high alloy temperature on the thin base HBT. An excessive alloy temperature of the collector metal will induce a high offset voltage ( $V_{\text{offset}}$ ), while an insufficient alloy temperature will be bad for the formation of ohmic contact to the collector, and a bad contact will lead to a high specific contact resistance ( $R_{\text{contact}}$ ). An HBT's performance is greatly affected under both conditions<sup>[7-9]</sup>.

In this paper, the alloy temperature dependence of  $V_{\text{offset}}$  and  $R_{\text{contact}}$  has been studied. In addition, the reason for the high  $V_{\text{offset}}$  at high alloy temperature is interpreted using Schottky clamped theory. The lower  $V_{\text{offset}}$  in our U-shaped emitter HBT than that in traditional strip emitter HBTs is explained by a formula deduced from the fundamental

device equation.

## 2 Experiment

Epitaxial structure plays a key role in an HBT's performance. A 100mm GaAs-substrate epitaxial wafer grown by MBE was provided by the Shanghai Institute of Microsystem and Information Technology of the Chinese Academy of Sciences. The structure of the epitaxial layers is shown in Table 1. A high concentration of indium in the 9th layer helps form a good ohmic contact between the emitter metal and epitaxy layer. The 4th layer, as a base layer, is designed with a thickness of only 40nm to enhance  $f_t$ , whereas the normal base layer thickness is over 70nm.

Table 1 Structure of the epitaxial layers

Layer No.	Composition $x$		Thickness / nm	Doping / $\text{cm}^{-3}$	Dopant
9	$\text{In}_x\text{Ga}_{1-x}\text{As}$	0.6	50	$>1 \times 10^{19}$	Si
8	$\text{In}_x\text{Ga}_{1-x}\text{As}$	0.6~0	50	$>1 \times 10^{19}$	Si
7	GaAs		120	$5 \times 10^{18}$	Si
6	$\text{In}_x\text{Ga}_{1-x}\text{P}$	0.5	50	$3 \times 10^{17}$	Si
5	GaAs		3	Undoped	
4	GaAs		40	$4 \times 10^{19}$	Be
3	GaAs		2	Undoped	
2	GaAs		150	$2 \times 10^{16}$	Si
1	GaAs		500	$5 \times 10^{18}$	Si
SI GaAs substrate					

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In the device fabrication, the key process involved a self-aligned U-shaped emitter, a lateral etched undercut (LEU), a high temperature alloy, etc.

In the self-aligned emitter, the emitter metal acts as the mask for etching the emitter and as a shield in depositing the base metal.

The majority of the extrinsic base-collector junction area is removed by LEU<sup>[10]</sup> with a selective etchant. This means that  $C_{bc}$ , a main nonlinear factor in HBTs, is greatly reduced in U-shaped emitter HBTs, compared to traditional strip emitter HBTs<sup>[11]</sup>. The profiles of U-shaped emitter and traditional strip emitter HBTs are shown in Fig. 1.

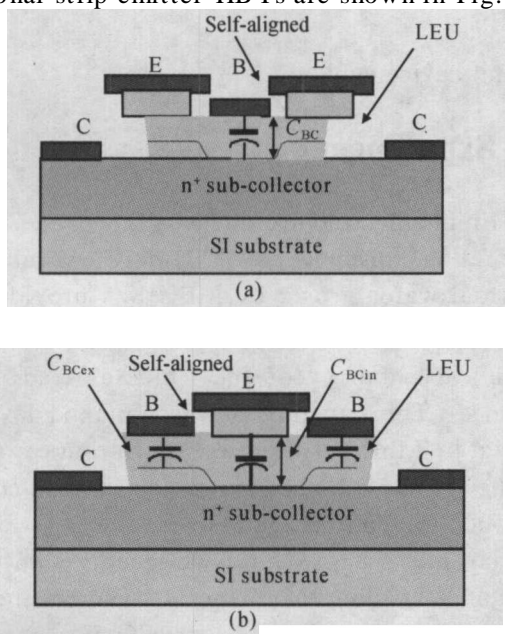


Fig. 1 Comparative profiles of U-shaped emitter and traditional strip emitter HBTs (a) U-shaped emitter; (b) Strip emitter

To find a proper alloy temperature for the thin base HBT, an experiment is conducted in a temperature range from 360 to 420 °C, and  $V_{\text{offset}}$  and  $R_c$  are measured.

### 3 Results and discussion

The ohmic contact resistance is measured using the transmission line method. The results are shown in Fig. 2. The specific contact resistance of the metal-semiconductor contact ( $R_{\text{contact}}$ ) is about  $6 \times 10^{-6} \Omega \cdot \text{cm}^2$ .  $R_{\text{contact}}$  varies slowly with temperature, but it increases faster as the temperature falls below 390 °C, and almost reaches  $1 \times 10^{-5} \Omega \cdot \text{cm}^2$  at

360 °C. Therefore, we think the alloy should be carried out above 390 °C.

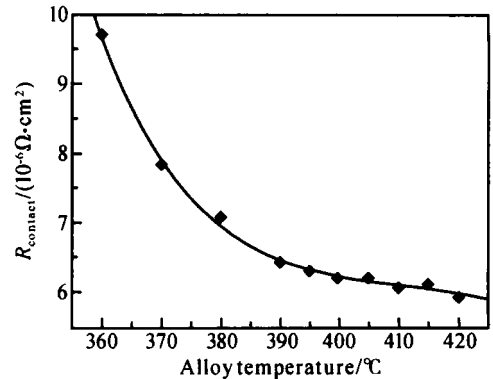


Fig. 2 Alloy temperature dependence of  $R_{\text{contact}}$

The alloy temperature dependence of  $V_{\text{offset}}$  is shown in Fig. 3. The symbols are the experimental results, and the solid line is the fitting curve.  $V_{\text{offset}}$  increases slowly from 0.085 to 0.09V when the alloy temperature increases from 360 to 390 °C. But when the temperature exceeds 400 °C,  $V_{\text{offset}}$  increases dramatically, and reaches a very high value, 0.32V, in 420 °C.

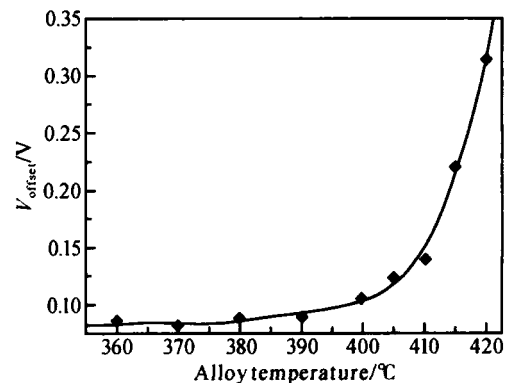


Fig. 3 Alloy temperature dependence of  $V_{\text{offset}}$  from 360 to 420 °C

First, from the experimental results shown above, we can conclude that the optimal alloy temperature for a thin base HBT is between 390 and 400 °C, for the best trade-off between  $V_{\text{offset}}$  and  $R_c$ .

Second, the high  $V_{\text{offset}}$  can be understood with Schottky clamped theory<sup>[12]</sup>. In a high temperature alloy, the base contact metal might penetrate the base layer and reach the collector area. Then a Schottky diode is formed between the base and the collector. As shown in Fig. 4, a Schottky diode is parallel to the normal base-collector p-n junction. Then the transistor is said to be "Schottky

clamped”. Since the turn-on voltage of a Schottky diode is lower than that of the p-n junction,  $V_{\text{offset}}$  is larger when Schottky clamping occurs. As  $V_{\text{offset}}$  and hence  $V_{\text{knee}}$  increase, the output power and power efficiency of the device decrease, which are harmful to the power amplifier with a low power supply.

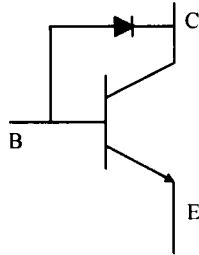


Fig. 4 Schottky clamped HBT

Another characteristic of the U-shaped HBT is that the value of  $V_{\text{offset}}$  is only about 0.09V, while that of a traditional strip emitter HBT is about 0.15V or higher<sup>[11]</sup>. InGaP/ GaAs HBTs have received much attention due to their inherent advantages, including better DC and AC performance, noise figure properties, and reliability than AlGaAs/ GaAs HBTs. However, a disadvantage of InGaP/ GaAs HBT is its larger offset voltage<sup>[9]</sup>. The lower  $V_{\text{offset}}$  of the U-shaped emitter HBT is attributed to the U-shaped layout structure and LEU technology. As can be seen in Fig. 1, the effective base-collector area of the U-shaped emitter HBT is greatly reduced. Base-collector junction capacitors ( $C_{bc}$ ) measured using ICCAP also demonstrate a reduction in U-shaped HBTs. Liu<sup>[12]</sup> studied the relation between  $V_{\text{offset}}$  and the area ratio  $A_c/A_e$  for graded HBTs, where  $A_c$  and  $A_e$  are the collector and emitter areas, respectively. But abrupt HBTs have not been discussed. From fundamental physical relations,  $V_{\text{offset}}$  can be computed by<sup>[12]</sup>

$$V_{\text{offset}} = \frac{-bc}{q} \frac{kT}{I_{cs}} \ln\left(\frac{I_{cs}}{f I_{es}}\right) + \frac{-bc}{be} I_b R_c + (1 - \frac{-bc}{be}) (V_{be} - I_b R_b) \tag{1}$$

where  $I_{cs}$  and  $I_{es}$  are the base-collector diode saturation current and base-emitter diode saturation current, respectively.  $\frac{-bc}{be}$  and  $\frac{-bc}{bc}$  are the ideality factors of the emitter current in the reverse-active mode and of the collector current in the forward-active mode, respectively, and  $f$  is the forward current transfer ratio. The first term of Eq. (1) is dominant in determining  $V_{\text{offset}}$ . Since the  $f$ ,  $I_{cs}$ ,

and  $I_{es}$  are defined as follows :

$$I_{es} = \frac{qp_{c0} A_c D_{pc}}{L_E} + \frac{qn_{b0} A_c D_{nb}}{L_B} \coth(X_b/L_b) \tag{2}$$

$$I_{cs} = \frac{qp_{c0} A_c D_{pc}}{L_C} + \frac{qn_{b0} A_c D_{nb}}{L_B} \coth(X_b/L_b) \tag{3}$$

$$f = \frac{1}{I_{es}} \times \frac{qn_{b0} A_c D_{nb}}{L_b} \coth(X_b/L_b) \tag{4}$$

From the above four equations, we get :

$$V_{\text{offset}} = \frac{-bc}{q} \frac{kT}{I_{cs}} \ln\left(1 + \frac{A_c}{A_e} \times \frac{L_b p_{c0} D_{pc}}{L_c n_{b0} D_{nb}}\right) + \frac{-bc}{be} I_b R_c + (1 - \frac{-bc}{be}) (V_{be} - I_b R_b) \tag{5}$$

$$L_c = \sqrt{D_{pc} \tau_c} \tag{6}$$

$$L_b = \sqrt{D_{nb} \tau_b} \tag{7}$$

$$D_n = \frac{kT}{e} \mu \tag{8}$$

where  $L_b$  is the minority electron diffusion length in the base and  $L_c$  is the minority hole diffusion length in the collector. Equation (5) shows that  $V_{\text{offset}}$  increases with  $A_c/A_e$ . Since the other parameters in Eq. (5) are mainly determined by epitaxial material,  $A_c/A_e$ , which is determined by layout, is the only variable. Calculations using Eq. (5) show a rough difference of 0.05V between the U-shaped emitter HBT and the traditional strip emitter HBT, which roughly agrees with the experimental results. There is still some error between the computations and experimental results, which may be caused by the fabrication process of the HBT.

### 4 Conclusion

For the best trade-off between  $V_{\text{offset}}$  and  $R_{\text{contact}}$ , a temperature in the range from 390 to 400, is appropriate for the alloy of thin base HBTs. High  $V_{\text{offset}}$  at high alloy temperatures can be interpreted using Schottky clamped theory. The lower  $V_{\text{offset}}$  in our U-shaped emitter HBT than in traditional strip emitter HBTs is attributed to the lower  $A_c/A_e$ . This further demonstrates the superiority of U-shaped emitters, which was shown by Bai *et al.*<sup>[10]</sup> with high  $f_t$  and  $f_{\text{max}}$ .

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## 合金温度对薄基区 InGaP/GaAs 残余电压和欧姆接触的影响\*

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**摘要:** 研究了薄基区 HBT 合金温度对残余电压  $V_{\text{offset}}$  和欧姆接触电阻  $R_{\text{contact}}$  的影响, 给出了薄基区 HBT 的最佳合金温度区域. 用肖特基钳位理论解释了合金温度过高导致  $V_{\text{offset}}$  偏大的现象. 从晶体管基本物理机制推导出  $V_{\text{offset}}$  与集电极、发射极面积比  $A_c/A_e$  的关系, 并用此解释了 U 形发射极 HBT 具有较小  $A_c/A_e$  的原因, 进一步证明了 U 型发射极结构的优越性.

**关键词:** 异质结双极型晶体管; U 形发射极; 合金; 残余电压

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